

# Issues associated with polarisation independence in silicon photonics

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## ABSTRACT

Silicon Photonics is experiencing a dramatic increase in interest due to emerging applications areas and several high profile successes in device and technology development. Despite early work dating back to the mid 1980s, dramatic progress has been made in recent years. Whilst many approaches to research have been developed, the striking difference between the work of the early to mid 1990s, and more recent work, is that the latter has been associated with a trend to reduce the cross sectional dimensions of the waveguides that form the devices. The question arises therefore as to whether one should move to very small strip waveguides (silicon wires) of the order of 250nm in height and a few hundred nanometres in width for improved device

performance but with little hope of polarisation independence, or to utilise slightly larger rib waveguides that offer more opportunity to control the polarisation dependence of the devices. In this paper we discuss devices suitable for one approach or the other and present designs associated both with strip and rib waveguides. In particular, we present designs of polarisation independent ring resonators with FSRs up to 12nm, we propose modulators for bandwidths in the 10s of GHz regime, and we present grating based couplers for rib and strip waveguides, and/or for wafer scale testing, as well as a novel means of developing Bragg gratings via ion implantation.

*Index Terms: Silicon-On-Insulator (SOI), rib waveguides, single mode condition, polarisation independence, optical modulators, Bragg gratings, grating couplers, ring resonators*

## I. INTRODUCTION

Silicon Photonics is a research field that is surprisingly mature in some senses, having been studied since the mid 1980s. In other ways, it is in its infancy, with some major advances being reported only very recently. The first waveguides were reported in the mid 1980s, in silicon on doped silicon [1], silicon on sapphire [2], silicon germanium [3], and Silicon on Insulator (SOI) [4,5]. The silicon on insulator platform, first reported in 1989, has by far, become the most popular of the four waveguide systems, and it is this platform that has formed the core of the work carried out by the Silicon Photonics Group at the University of Surrey, which was also established in 1989. The first results from the group were published in early 1991, demonstrating waveguides formed by the SIMOX process [6]. These early waveguides

exhibited losses as high as 30dB/cm, but within a year the same group had reported waveguides with a loss of less than 1dB/cm, demonstrating the viability of the technology [7].

The first silicon based optical modulators were proposed in 1986, with modelling suggesting a  $\pi$ -radian phase shift could be achieved in a device less than a 1mm long. The corresponding loss was less than 1 dB at  $\lambda = 1.3\mu\text{m}$  for both TE and TM polarisations [8]. However, the electrical power densities required to drive early modulators was very high, and it was not until 1993 that the Surrey group proposed a 3-terminal device that reduced power consumption by an order of magnitude [9, 10]. Device variants had bandwidths of up to 20MHz, and were based upon a waveguide with large cross sectional dimensions of the order of  $6\mu\text{m}$ , and a drive current of only 7mA [11]. Nevertheless more recent work has proposed and/or demonstrated fast modulators in smaller waveguides, with drive currents below 1mA. Some of these devices will be discussed later, together with our other work on ring resonators, grating based devices and optical couplers. Firstly however, let us consider some fundamental issues associated with shrinking the waveguide cross sectional dimensions.

## II. FUNDAMENTAL WAVEGUIDE ISSUES

Single-mode SOI rib waveguides with large cross section have been studied extensively by a number of researchers [12-17] to find single mode behaviour at the same time as low propagation loss. The majority of these photonic devices in SOI have been studied in waveguides that are multi-micron in cross sectional dimensions (of the order of  $5\mu\text{m}$ ), to facilitate low-loss coupling to and from optical fibres. Soref *et al.* [13] first proposed a simple expression for these large ribs waveguides, related to their geometry to ensure that they satisfied the single-mode condition (SMC):

$$\frac{W}{H} \leq \alpha + \frac{r}{\sqrt{1-r^2}}, \text{ for } 0.5 \leq r \leq 1 \quad (1)$$

where  $r$  is the ratio of slab height to overall rib height,  $W/H$  is the ratio of waveguide width to overall rib height, and  $\alpha = 0.3$ . The analysis of the waveguides was limited to shallow etched ribs ( $r > 0.5$ ), and hence deeply etched rib waveguides were not considered. Furthermore, the waveguide dimensions were assumed to be larger than the operating wavelength. Their analysis was based on the assumption that high order vertical modes (i.e. modes other than the fundamental mode) confined under the rib waveguides, were coupled to the outer slab region during propagation, therefore yielding high propagation losses for the higher order modes. Thus the waveguides behave as single mode waveguides, as all other modes are lost. Other authors have also considered the single mode for large waveguides, and produced similar expressions (e.g. [14], [15]). However, the current trend in silicon photonic circuits to move to smaller device dimensions for improved cost efficiency and device performance can come at some cost to other performance parameters, notably in the polarisation dependence of the circuits if they are not carefully designed. Furthermore, Soref's design equation (equation 1), cannot be applied to small and deeply etched rib waveguides. In order to maintain the consistency with other work [13, 15-17], we have used the full-vectorial beam propagation method (BPM) [18] to analyse the deeply etched rib waveguide structure, mode propagation within it, and polarisation independence. Some results of the SMC condition have also been verified by the finite element method (FEM) [19]. For rib waveguides we have evaluated the single mode cut-off condition by determining when the first mode of higher order than the fundamental mode begins to propagate. Because the waveguides are relatively small, for high etch depths ( $r < 0.5$ ), the single mode condition

becomes dominated by boundary conditions, and hence the conditions for quasi-TE and quasi-TM modes begin to diverge. We have also determined the difference between the effective indices of the fundamental mode as a function of waveguide width and etch depth (and hence parameter  $r$  defined above), for a given waveguide height. The condition when the effective indices are equal is defined as the zero birefringence condition, and there are up to two such events for each waveguide etch depth. By determining these conditions we can plot a “zero birefringence locus” for each waveguide height [20].

The simulations were carried out for rib waveguides of silicon ( $n_g=3.477$ ) on silica ( $n_s=1.444$ ) and an upper cladding that is air ( $n_c=1$ ), although it is a simple task to extend the work to an upper oxide cladding. SOI rib waveguides with an overall height in the range  $H = 1.00\mu\text{m}$ , to  $1.50\mu\text{m}$  were analysed at a wavelength of  $1.55\mu\text{m}$ . Both the single mode condition and the zero birefringence conditions can conveniently be plotted on the same curve to determine the waveguide parameters that allow both conditions to be satisfied simultaneously. For example, Figure 1 shows such plots for waveguide heights of  $1.35\mu\text{m}$  (Figure 1(a)) and  $1.5\mu\text{m}$  (Figure 1(b)). It can be seen from Figure 1 that for truly single mode behaviour, it is the quasi TM condition that is the limiting condition, because if this is satisfied, then the quasi TE condition is automatically satisfied. Consequently, for both single mode behaviour and polarisation independence, then the waveguide design should lie on the zero birefringence locus, below the quasi-TM single mode boundary, in the bottom right hand corner of figures 1(a) and 1(b). From this and other data [20], [21], we can extract design rules to aid the design of single mode rib waveguides, for waveguide heights in the range  $H = 1.00\mu\text{m}$  to  $1.50\mu\text{m}$ .

$$\frac{W}{H} \leq 0.05 + \frac{(0.94 + 0.25H)r}{\sqrt{1-r^2}} \text{ for } r \leq 0.5 \text{ and } 1.0 \leq H \leq 1.5 \quad (2)$$

$$D_{\min} = 0.06 \times 10^{-6} + 0.556H \quad (3)$$

Equation 2 defines the quasi-TM single mode boundary, and hence provides guidance on the geometric limitations to retain single mode behaviour, whilst equation 3 defines the minimum etch depth required to obtain polarisation independence.

### III. RING RESONATORS IN RIB WAVEGUIDES

We can utilise the information of the previous section to design a polarisation independent ring resonator. Such a resonator requires polarisation independent waveguides for matching of the TE/TM phase shifts around the ring waveguides, but also requires polarisation independent directional couplers to transfer light to and from the ring. A directional coupler comprises two waveguides in close proximity such that the evanescent fields of the optical modes overlap, and light can transfer from one waveguide to the other. We have previously described how to design such a coupler [22], by allowing multiple passes of light from one of the waveguides in the directional coupler to the other. For example Figure 2(a) shows modelling of the optical power in the two arms of a directional coupler. For the purposes of the modelling, the left arm of the coupler was excited with an optical field. With propagation distance  $z$ , light transfers from the left waveguide to the right, and back again in a cyclic manner. The goal of the modelling is to find a coupler length that yields the same degree of power transfer from the left waveguide to the right, for both polarisations.

Figure 2(a) demonstrates that, power transfers in a shorter length for the TE polarisation than for TM. However for three transitions of the TM mode, and 5 transitions of the TE mode, the power in the left waveguide has transferred to the right waveguide, over a coupler length of approximately  $500\mu\text{m}$ , and hence for this fixed coupling length the device is polarisation independent. A ring resonator utilising this principle was fabricated, and the experimental characteristics are shown in Figure 2(b), which clearly demonstrates polarisation independent performance over a spectral width of three times the free spectral range (FSR). This device was based upon a waveguide height of  $1.35\mu\text{m}$  and a rib width of  $0.8\mu\text{m}$ .

However, the resultant FSR is very small, due to the relatively large circumference of the ring resonator, which in turn is a direct result of the size of the waveguides. Consequently it could be argued that resonators based upon strip waveguides would be more useful as the ring circumference could be much smaller ( $\sim 5\mu\text{m}$ ) leading to FSRs of the order of 10-20 nm (e.g. [23]). However, we can push the large ring resonators further either by simply shrinking the ring circumference, or by employing cascaded or multiple ring resonators. Let us first consider shrinking the ring circumference. The device resulting in the data of Figure 2(b) had a bend radius of  $400\mu\text{m}$  and a total circumference of  $3513\mu\text{m}$ , resulting in an FSR of only 193pm. We have also fabricated a series of other resonators with varying circumference, based upon identical waveguide dimensions. The results of these devices are summarised in Table 1. This table shows that if the bend radius is reduced to  $25\mu\text{m}$ , the FSR can be increased significantly to 815pm. However, this is still very small for many applications. Consequently we can investigate both cascaded resonators and devices based upon serial coupling of multiple rings.

We have carried out modelling of both such devices, based upon the experimental results summarised in Table 1. For example if we consider cascaded ring resonators based upon our devices with bend radii of 25 and 50 $\mu\text{m}$ , the data of Figure 3(a) results, exhibiting an FSR of approximately 4.1nm. In such a configuration light propagates through the first resonator, straight section and finally through the second resonator towards the output drop port. If rings are different, resonant conditions are different and only wavelengths satisfying both of them will be present at the output. Therefore, the device acts as “and” function between two rings and the net transfer function (Figure 3(a)) can be approximated by multiplying responses of single racetrack resonators employing 310 $\mu\text{m}$ -couplers and bend radii of 25 $\mu\text{m}$  and 50 $\mu\text{m}$  assuming the same coupling and propagation conditions as for single stage filters. Alternatively, a device with serially coupled rings of radii of 25 and 33.24 $\mu\text{m}$  and with the length of the coupler of 310 $\mu\text{m}$ , exhibiting the so called Vernier effect, exhibit a much improved FSR of approximately 12nm. The value for the second radius has also been obtained by observing single-stage filter response for 25 $\mu\text{m}$  racetrack, assuming the same coupling conditions and taking into account Vernier condition  $m \times \text{FSR}_1 = n \times \text{FSR}_2$  [24]. By choosing these values for bend radii and the length of the coupler, the total FSR becomes 15 times bigger than the FSR for 25 $\mu\text{m}$ -single ring, the spectrum of which is given in (Figure 3(b)), but it should be also noted that this design requires very careful matching of the coupling coefficients and, consequently, control of fabrication issues. This figure could be improved further by reducing the coupler length and hence the circumference by using MMI couplers. Nevertheless it is clear that FSRs are possible in polarisation independent resonators that approach the FSR performance of very small ring resonators that would almost certainly be highly polarisation dependent.

It is worth considering one further improvement of the multiple ringed resonator structure. By serially coupling similar rings (rather than dissimilar rings for FSR improvement), we can vary the FWHM of the resonance, and hence change the effective Q factor of the device. For example Figure 4 shows experimental results of ring resonator devices with single, double and triple ringed devices, each with rings of the same size. The devices were based on the same waveguides as previously, and a directional coupler of  $2010\mu\text{m}$ , resulting in a full width half maximum (FWHM) of  $27\text{pm}$  for a single ring device,  $22\text{pm}$  for a dual ring device and  $17\text{pm}$  for a triple ring device. Figure 4b shows that the experimental results of a double ring resonator compares extremely well with the theoretical transfer function, demonstrating that we have excellent control of the design parameters of such devices.

#### IV. GRATING BASED DEVICES

Gratings have numerous applications in optical circuits including filtering, coupling to/from circuits, and phase matching in directional couplers, and as Bragg reflectors. Since gratings are highly polarisation dependent devices, it is questionable whether there are applications in polarisation independent waveguides. However, if the grating can be made to respond only to one polarisation, then cascading two gratings can result in a device that is effectively polarisation independent. A Bragg reflector is an obvious example of such a device.

We have carried out grating work in silicon waveguides ranging in height from  $1\mu\text{m}$  to  $1.5\mu\text{m}$ . Our earliest work dates back to 1998, and was based upon grating couplers for input/output coupling. Figure 5 shows grating couplers on the surface of  $1\mu\text{m}$  waveguides. Figure 5a shows

a conventional castellated grating that exhibited a measured out-coupling efficiency of 72% at a wavelength of  $1.3\mu\text{m}$  [25], and Figure 5b shows a blazed grating also on a  $1\mu\text{m}$  waveguide that exhibited an increased out-coupling efficiency of 84% at a wavelength of  $1.3\mu\text{m}$  [26]. Such gratings are useful for lab based coupling experiments, but may also have much wider application in wafer scale testing of silicon photonic circuits.

We are now working on similar gratings, as well as Bragg gratings, that can be formed by ion implantation of oxygen, or other species. The advantage of such an approach is that there is no physical etch of the waveguide surface, and planar surface retention is easier to process. Furthermore, they are more convenient for integration with other devices such as heaters to enable thermal tuning of the gratings, or perhaps with electronics. The development of gratings fabricated via ion implantation relies on a modification of the refractive index at the waveguide surface in a periodic manner. This can be achieved in silicon by ion implantation through a mask, and hence the design is critically dependent upon achieving good control of the surface implant. Thus modelling of the surface implant as well as the required subsequent annealing process is important. Grating depths up to 150nm are being considered. Oxygen is a good choice for the implant, as the annealing process results in a definitive interface between the silicon dioxide resulting from the oxygen implant, and the remaining surface silicon, similar to work reported by Bussman et al [27]. Figure 6 shows the predicted implant profile of 30keV oxygen into silicon. Clearly the implant is not truly a surface implant, but the subsequent annealing process results in diffusion of oxygen into a near stoichiometric surface layer of  $\text{SiO}_2$ .

The preceding examples of grating structures are compatible with rib waveguides of the order of 1-2 $\mu\text{m}$  in height. However, we have built upon our work on surface gratings to develop a coupler device for coupling into very small waveguides, such as silicon wires, for applications in which polarisation independence is less important. We have developed a device we have entitled the Dual Grating Assisted Directional Coupler (DGADC) [28].

A DGADC in SOI is shown in Fig. 7a [28]. A thick waveguide and two separation layers are fabricated in SiON for refractive index control over a broad range (1.45-2). The top layer is 5  $\mu\text{m}$  thick with refractive index close to the refractive index of optical fibre, resulting in an insertion loss  $\leq 0.05$  dB from the fibre to this first waveguide.

A fibre is butt coupled to the thick SiON waveguide and subsequently the light is coupled to a  $\text{Si}_3\text{N}_4$  waveguide using the first grating, and to the thin ( $\approx 1/4 \mu\text{m}$ ) SOI waveguide via the second grating. The silicon nitride waveguide is crucial for the operation of the device because it enables highly efficient coupling at both gratings, consequently forming an efficient DGADC. This waveguide bridges the gap between SiON and Si layers in both refractive index and thickness. The buried oxide layer serves as the lower cladding layer, for isolation from the substrate, hence removing any leakage loss towards the substrate. Figure 7b shows the theoretical efficiency of the structure, achieved via modelling. The total efficiency is expressed as the product of the efficiencies of the first grating ( $\eta_1$ ) and the second grating ( $\eta_2$ ), as a function of the thickness of the  $\text{Si}_3\text{N}_4$ . The efficiency is plotted for three different combinations of grating heights. The most important point to notice from this graph is that the theoretical coupling efficiency can exceed 90%. We have yet to fabricate the optimum device, but we have recently reported preliminary

results measured of a device with a surface waveguide height of  $\sim 3.7\mu\text{m}$ , which in turn means that the theoretical coupling is 60% [29], rather than  $>90\%$  in the optimized device. However, the experimentally determined efficiency was 55%, very close to the theoretical value. This is already a competitive performance for coupling to such small waveguides, but perhaps more importantly, this result means that the DGADC remains amongst the most promising devices for very high efficiency coupling to very small waveguides. The measured bandwidth of  $\sim 5\text{nm}$  can be broadened significantly by chirping and by varying the duty cycle of the gratings.

## V MODULATORS FOR RIB AND STRIP WAVEGUIDES

Due to the fact that fast modulators in silicon must probably be realized via the plasma dispersion effect, modulators will always be faster and more efficient in a smaller waveguide as compared to a similar modulator in a larger waveguide. However, one of the fastest modulators reported to date has been fabricated in a rib waveguide [30]. Therefore both rib and strip waveguides are worthy of consideration.

In 2003/4, we reported simulation of a *p-i-n* modulator based upon a rib waveguide with an overall waveguide height of approximately  $1\mu\text{m}$  [31, 32]. The structure is shown in Figure 8a. The doping profile of the  $n^+$  regions was optimised to provide maximum modulation speed. This optimum doping profile in the side  $n^+$  regions was achieved through modelling of a series of different implantation steps such that the peak concentration at the surface of the  $n^+$  contact was approximately  $10^{20}\text{cm}^{-3}$  and decreased to approximately  $10^{16}\text{cm}^{-3}$  at the interface between the silicon and the buried oxide. The determination of this profile was achieved through the process simulation package, ATHENA from SILVACO [33]. The optimum profile predicted a current

requirement to achieve a phase modulation of  $\pi$  radians,  $I_r$  of 0.7mA, as well as rise and fall times of 0.38ns and 0.13ns respectively, corresponding to a device bandwidth of the order of 1GHz. We also showed that by overdriving, the rise and fall times of the modulator can be further improved to provide a bandwidth in excess of 5GHz, an effect that has since been used by Xu et al [34] to overdrive a ring resonator modulator in a strip waveguide to achieve a data rate of 1.5Gbit/s.

We have recently reported modelling of a faster modulator suitable for width dimensions similar to those of a strip waveguide [35]. This device is shown in Figure 8b. It is fabricated in a waveguide that, whilst having dimensions approaching those of silicon wire dimensions, is technically still a rib waveguide, and hence some polarisation control is still possible. The device has an asymmetrical  $pn$  structure where two slab regions are joined as a common cathode and two poly-silicon regions are joined as a common anode. Both  $n^+$  and  $p^+$  regions were modelled as highly doped regions with peak doping concentrations of  $1 \times 10^{19} \text{cm}^{-3}$ . The structure is based around an overall silicon thickness of  $0.45 \mu\text{m}$ , etched rib waveguides  $0.415 \mu\text{m}$  wide with a slab thickness of  $0.1 \mu\text{m}$ . These dimensions were chosen to approach polarisation independence operational regime. The silicon slab and the bottom part of the rib have an  $n$ -type background doping concentration of  $4 \times 10^{17} \text{cm}^{-3}$  and the top part of the rib has a  $p$ -type background doping concentration of  $2 \times 10^{17} \text{cm}^{-3}$ . The  $n^+$  doped regions are situated on both sides of the waveguiding region, in the slab,  $1.5 \mu\text{m}$  from the centre of the waveguide. Furthermore the poly-silicon  $p^+$  doped regions are situated on both sides of the top of the rib in order to reduce the losses resulting from the poly-silicon and aluminium contacts.

The modulation mechanism was carrier depletion from the  $pn$  junction. Carrier losses induced were minimised in our design. Modulation was proposed by including two devices in a Mach Zehnder interferometer, to operate in push-pull mode. In the ‘off’ state, both devices were reverse biased to 5V such that no phase difference was induced between the arms of the Mach Zehnder. In the ‘on’ state, one arm was reverse biased to 10V whilst the bias on the other was reduced to 0V, to induce a  $\pi$  radian phase shift between the arms. The transient time of each device was characterised by ATLAS [33] in terms of carrier concentration against time. The carrier concentration profile for a reverse bias of 5 volts was taken as a reference. The change in carrier concentration compared to a reverse bias of 5 volts (against time) was converted to a change in refractive index profile (against time). This profile was then used in BeamPROP [18] simulation to determine the change in effective index, which was subsequently converted into a change in phase shift (with time).

The rise and fall times of the proposed device have both been calculated to be 7ps for a reverse bias of 5 volts, predicting an intrinsic bandwidth of several 10s of GHz, although in practice large peak currents will probably reduce this somewhat. This modulator provides improved performance in terms of loss and bandwidth compared to the strip waveguide device of Barrios and Lipson [36]. For a more detailed overview of modulators in silicon photonics, see for example [37].

## VI. CONCLUSION

The current trend in silicon photonics is to move to small waveguide dimensions, resulting in increased difficulty in maintaining single mode operation whilst simultaneously designing for polarisation independence. However, it is possible to achieve such a design for small rib

waveguides of the order of  $1.0 - 1.5\mu\text{m}$  in height. We have provided guidelines to aid such design.

We have demonstrated a polarisation independent design in ring resonators where such behaviour is particularly important. The intrinsic FSR of such devices is small due to the relatively large circumference of ring resonators based upon rib waveguides, and it is for this reason that some other authors have moved to smaller devices fabricated in strip waveguides. However, we have demonstrated that it is possible to achieve respectable FSRs in polarisation independent rib waveguides, with a predicted FSR based on our existing results being as large as  $12\text{nm}$ , and further improvement envisaged. We have also demonstrated additional control of the quality factor of such devices via serial cascading of multiple rings.

However, we also have an interest in strip waveguide work, and we have reported both grating based device designs and modulator designs in waveguides compatible with either strips or ribs. In particular the modulators offer the potential of multi tens of GHz modulation. In order to address the problem of coupling to very small waveguides we have designed and fabricated a DGADC for coupling to very small waveguides, promising coupling efficiencies up to 90%, with 55% reported to date.

The question to whether to pursue polarisation independence however, remains complex. In particular it is related to the application in question. For example, it is likely that for optimum modulation speed, a strip waveguide based modulator placed directly in front of a laser would not be required to exhibit polarisation independence because the laser is inherently polarised.

Alternatively, a modulator fed via an optical fibre is much more likely to be required to satisfy polarisation independent performance. Therefore we can envisage systems in which the performance of some devices is considerably more critical than others, and hence the question of whether to pursue polarisation independence must be viewed from the perspective of the application in question. Fortunately, the flexibility of silicon photonics means there is room for both approaches, and each approach has advantages in some application areas.

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## FIGURE CAPTIONS

Figure 1a The singlemode condition and the zero birefringence condition for a rib waveguide with height  $H = 1.35\mu\text{m}$ .

Figure 1b The singlemode condition and the zero birefringence condition for a rib waveguide with height  $H = 1.5\mu\text{m}$ .

Figure 2a Modelling of a directional coupler to show polarisation independent transfer of power from the left waveguide to the right waveguide, for a coupling length of  $500\mu\text{m}$ .

Figure 2b Measured spectral response of a polarisation independent racetrack resonator.

Figure 3a Modelled response of cascaded ring resonators based upon experimentally measured devices with bend radii of  $25\mu\text{m}$  and  $50\mu\text{m}$ , a coupler length of  $500\mu\text{m}$ , and waveguide height of  $1.35\mu\text{m}$ .

Figure 3b Modelled response of serially coupled double-ring resonators based upon experimentally measured devices with bend radii of  $25\mu\text{m}$  and  $50\mu\text{m}$ , a coupler length of  $210\mu\text{m}$ , and waveguide height of  $1.35\mu\text{m}$ .

Figure 4a Measured FWHMs of a Triple Ring Resonator (TRR), a Double Ring Resonator (DRR) and a Single Ring Resonator (SRR) with similar serially coupled rings are 17, 22 and 27pm, respectively.

Figure 4b Theoretical and measured responses of the Double Ring Resonator (DRR) with similar rings.

Figure 5a Cross section of a castellated surface grating coupler fabricated on a 1 $\mu$ m silicon waveguide. The period is approximately 400nm, designed to operate at  $\lambda = 1.3\mu$ m, and the output efficiency was 72%.

Figure 5b Cross section of a blazed surface grating coupler fabricated on a 1 $\mu$ m silicon waveguide. The period is approximately 383nm, designed to operate at  $\lambda = 1.3\mu$ m, and the output efficiency was 84%.

Figure 6 Simulation of 30keV Oxygen ions implanted into the surface of a silicon waveguide, showing the implanted profile and ion induced damage prior to annealing.

Figure 7a Cross sectional view of a Dual Grating Assisted Directional Coupler (DGADC), to couple light from an optical fibre to a small silicon waveguide

Figure 7b Predicted efficiency of the DGADC depicted in figure 7a for a range of combinations of grating heights

Figure. 8a Cross sectional view of a p-i-n phase modulator, designed for operation in a  $0.98\mu\text{m}$  height rib waveguide

Figure. 8b Cross sectional view of a depletion modulator for inclusion in a push-pull Mach Zehnder interferometer.

Table 1. A comparison of the FSR achieved for ring resonators fabricated with varying circumference.

Figure 1(a)

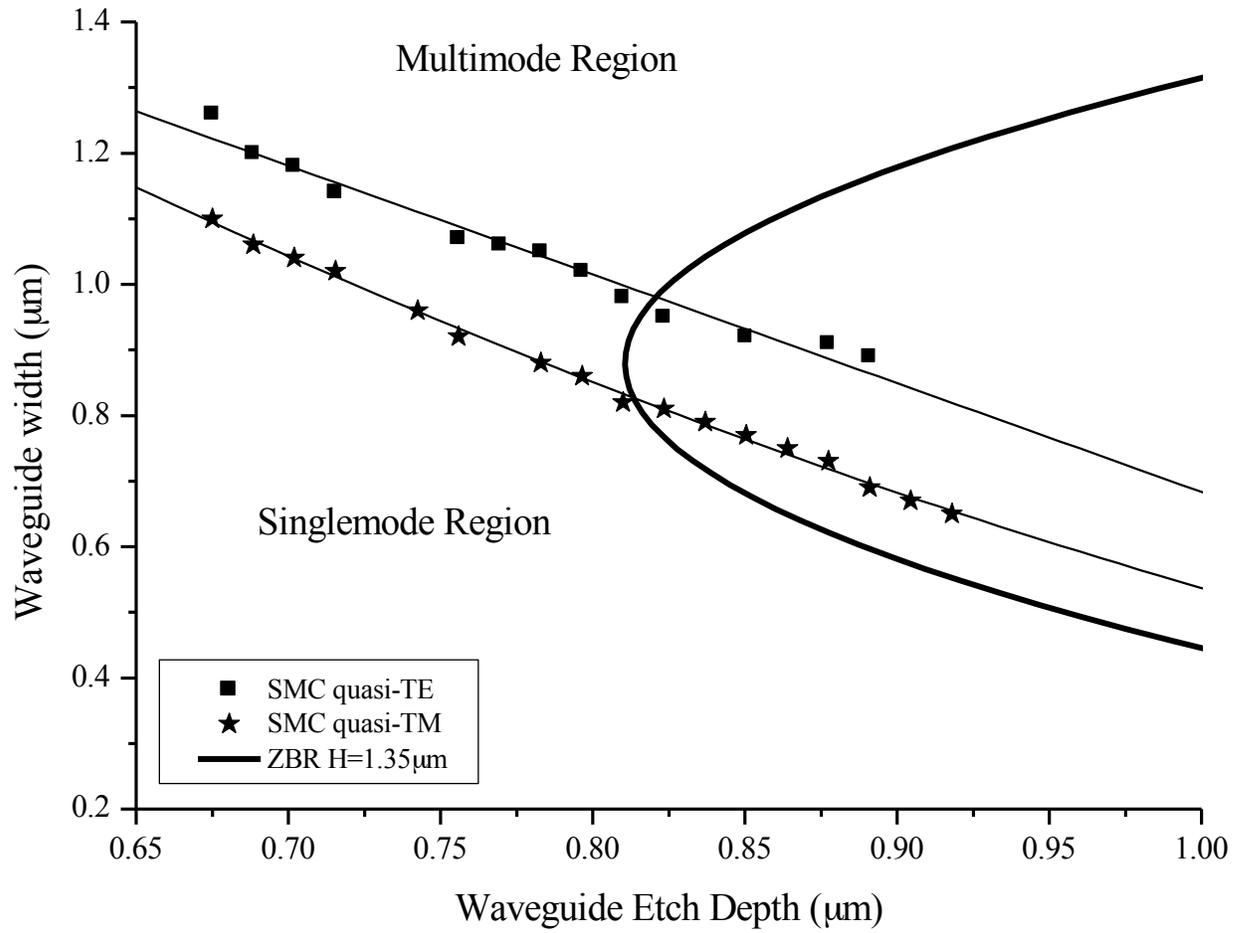


Figure 1(b)

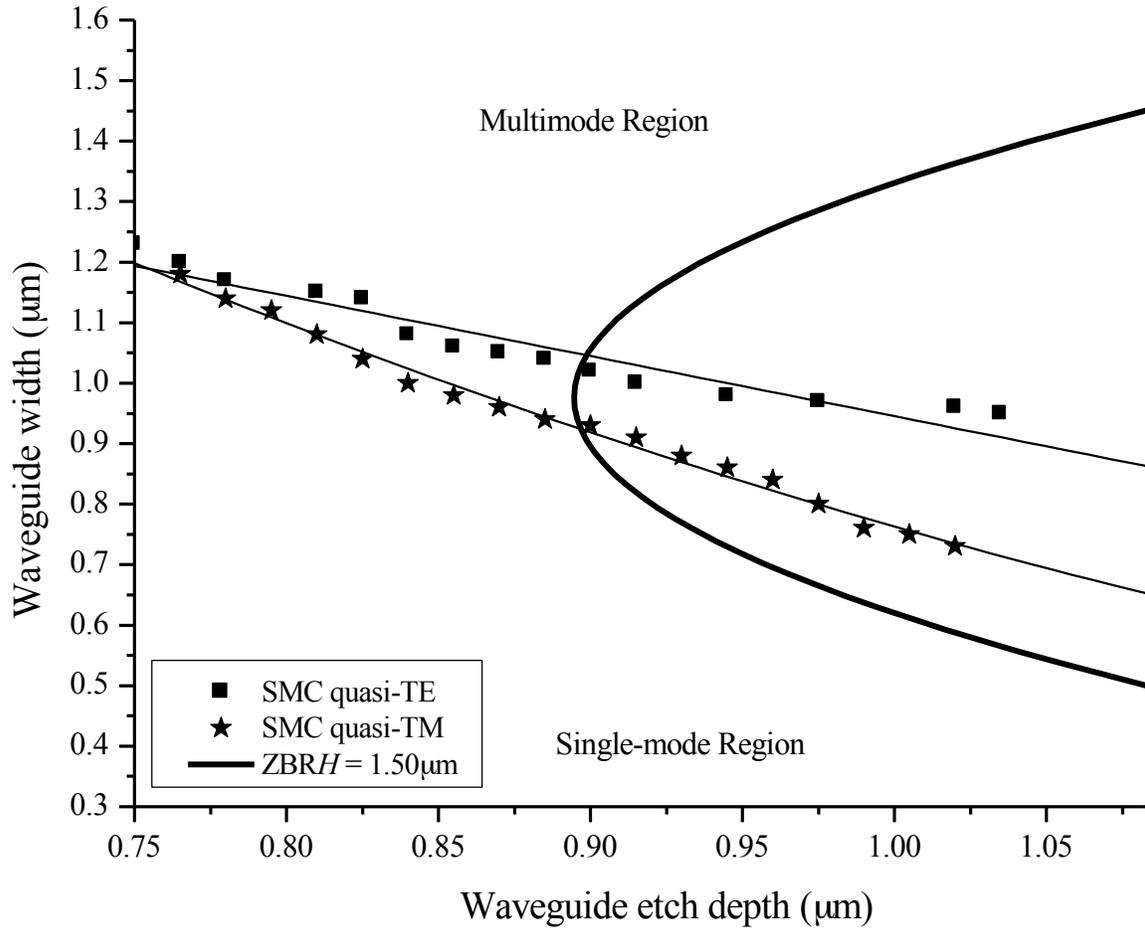
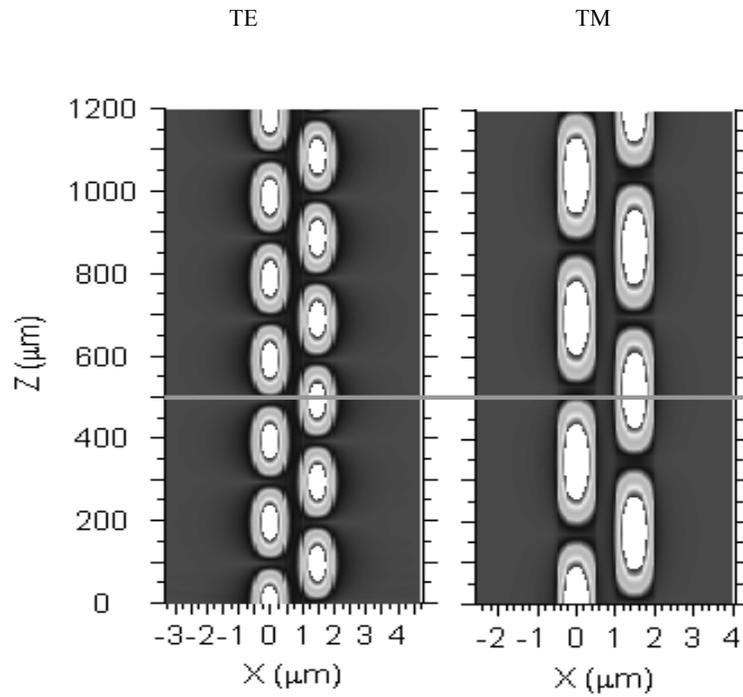


Figure 2(a)



(b)

Figure 2(b)

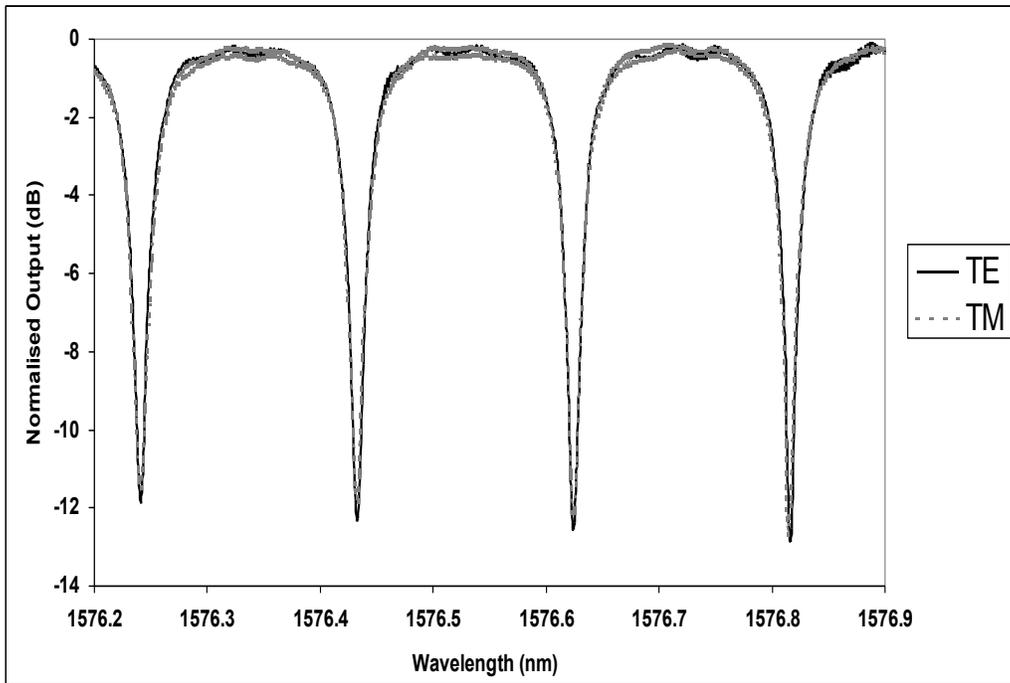


Figure 3a

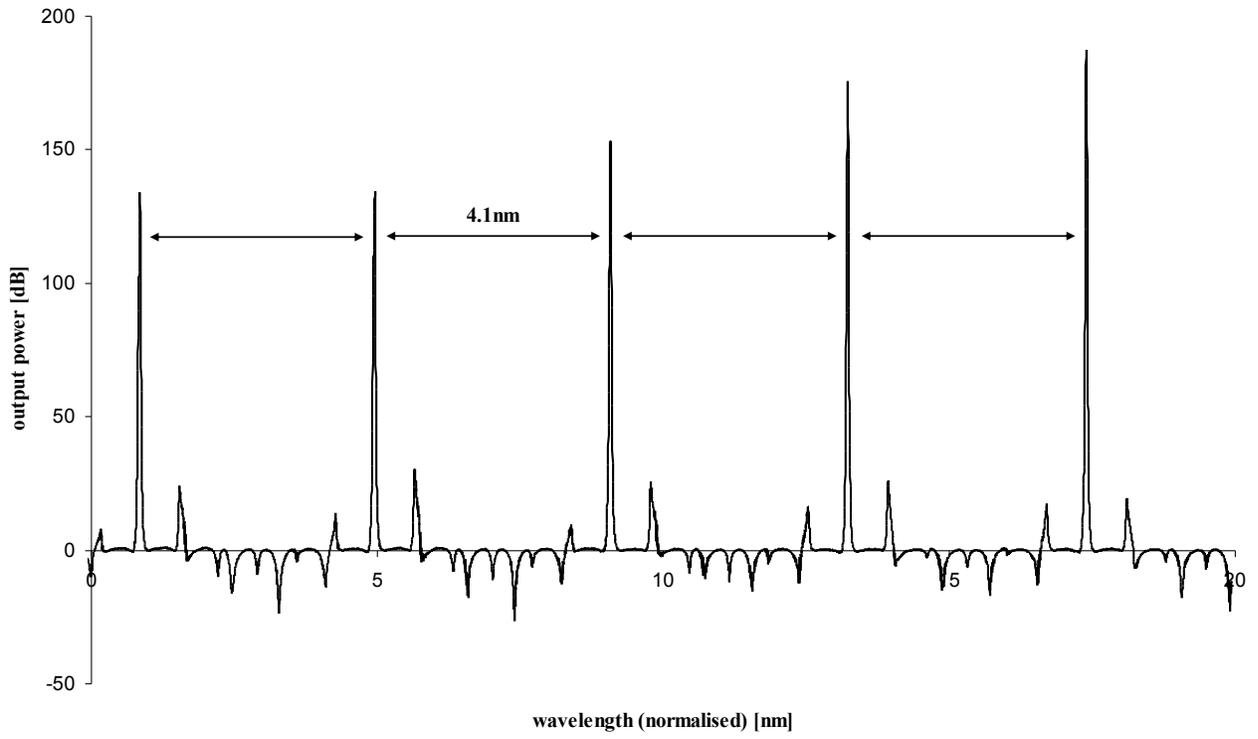


Figure 3(b)

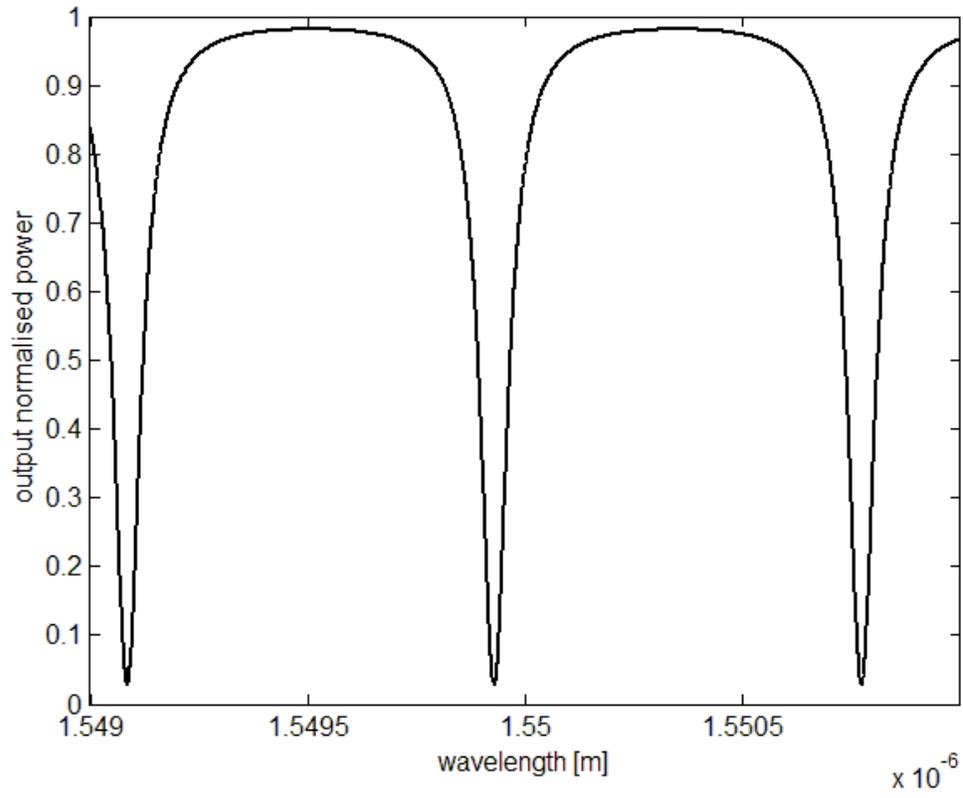


Figure 4a

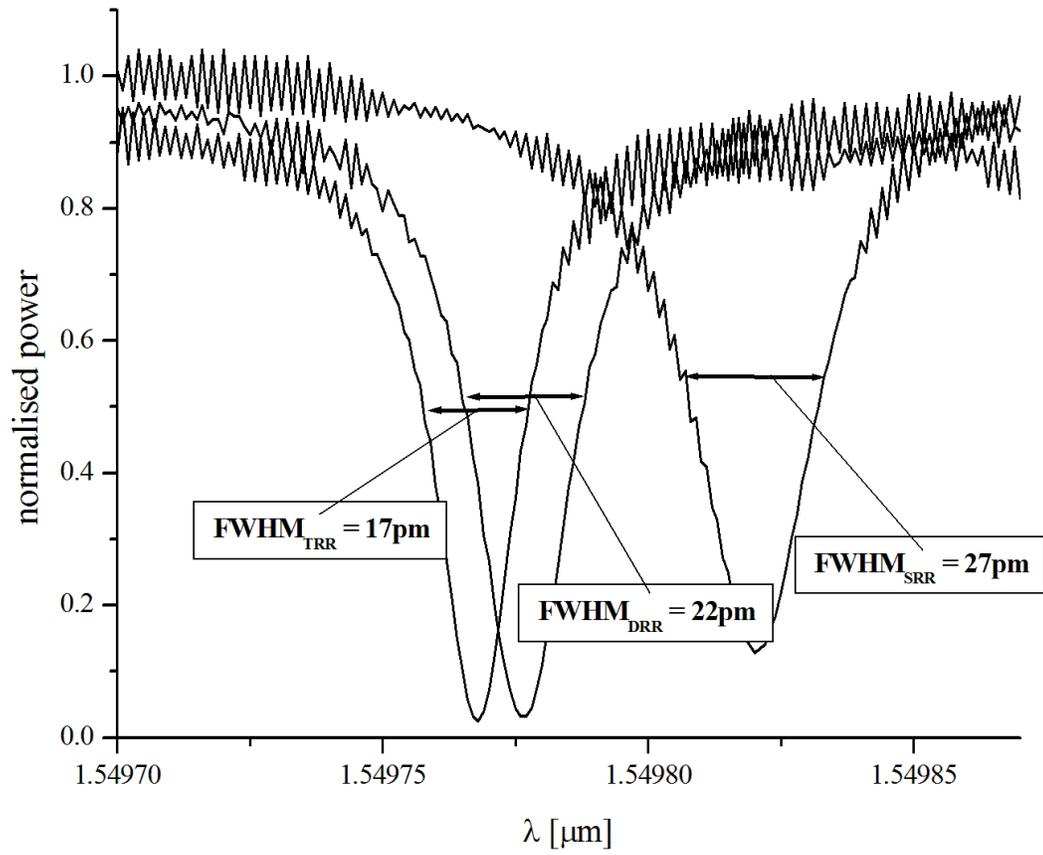


Figure 4b

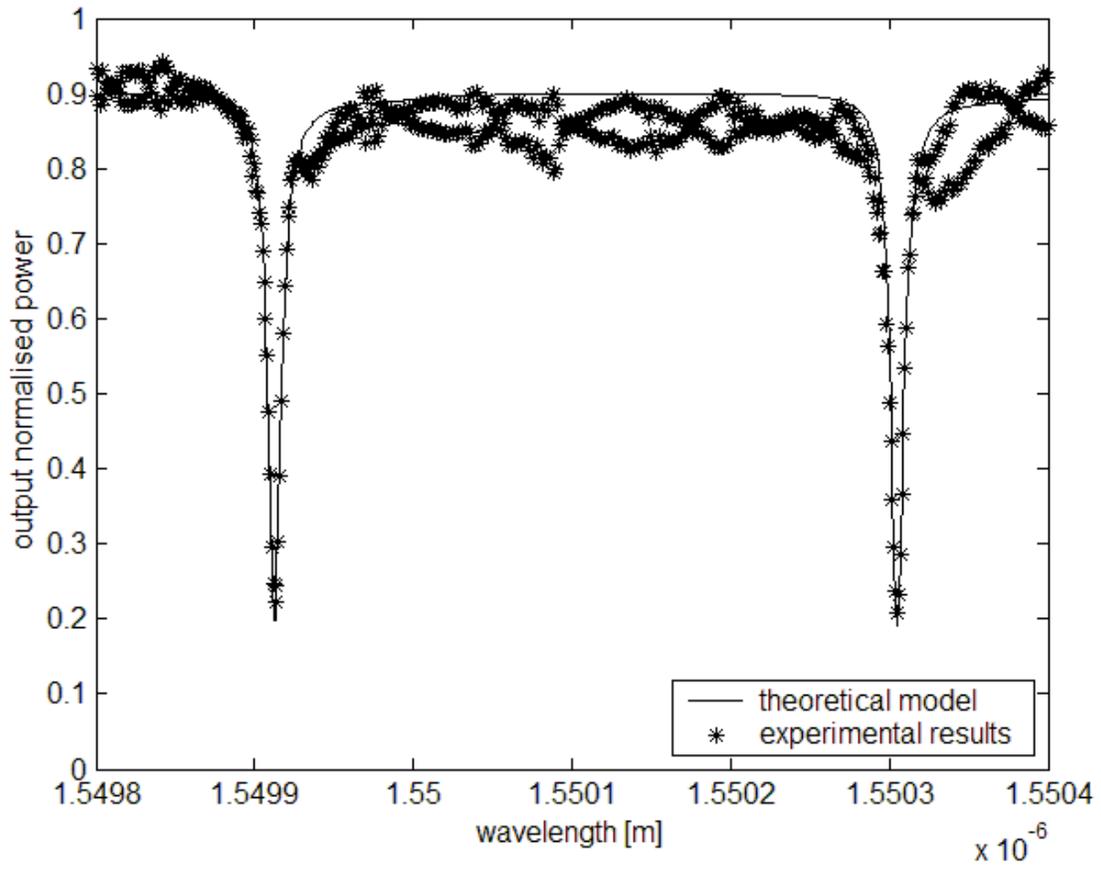


Figure 5a

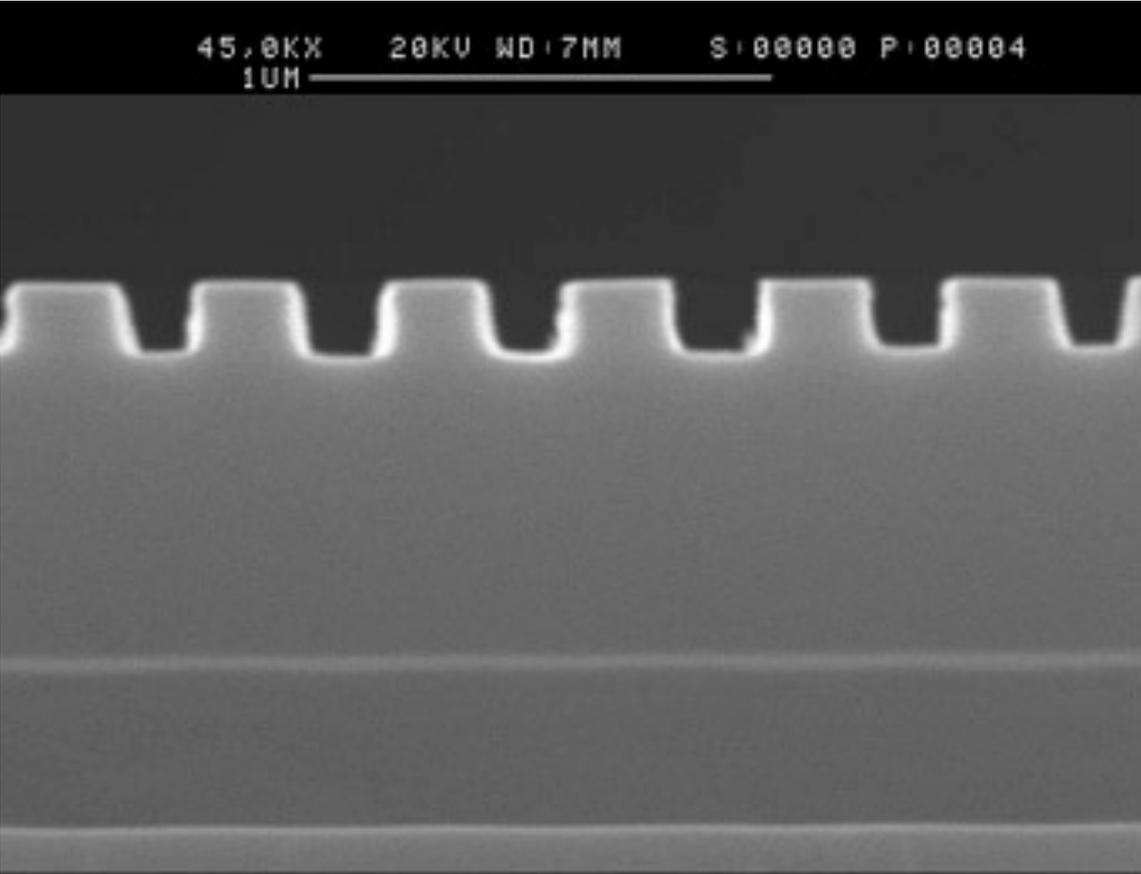


Figure 5b

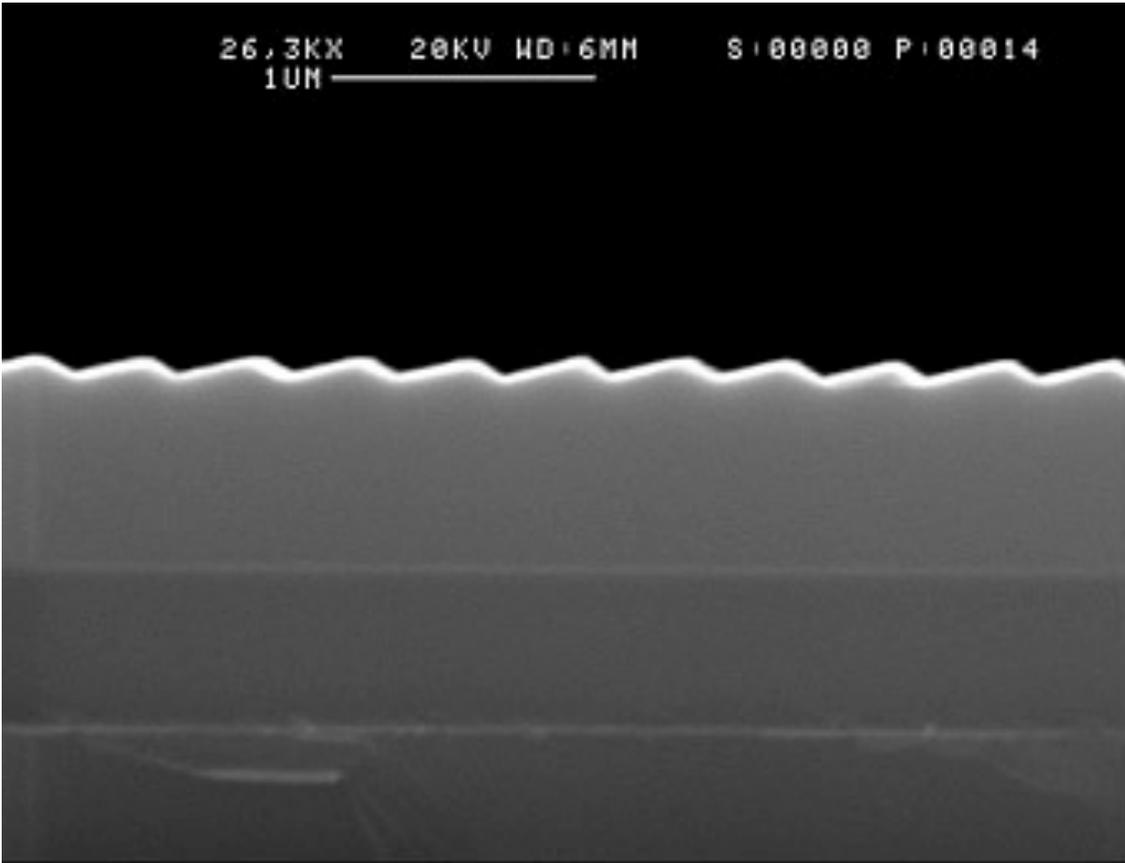


Figure 6

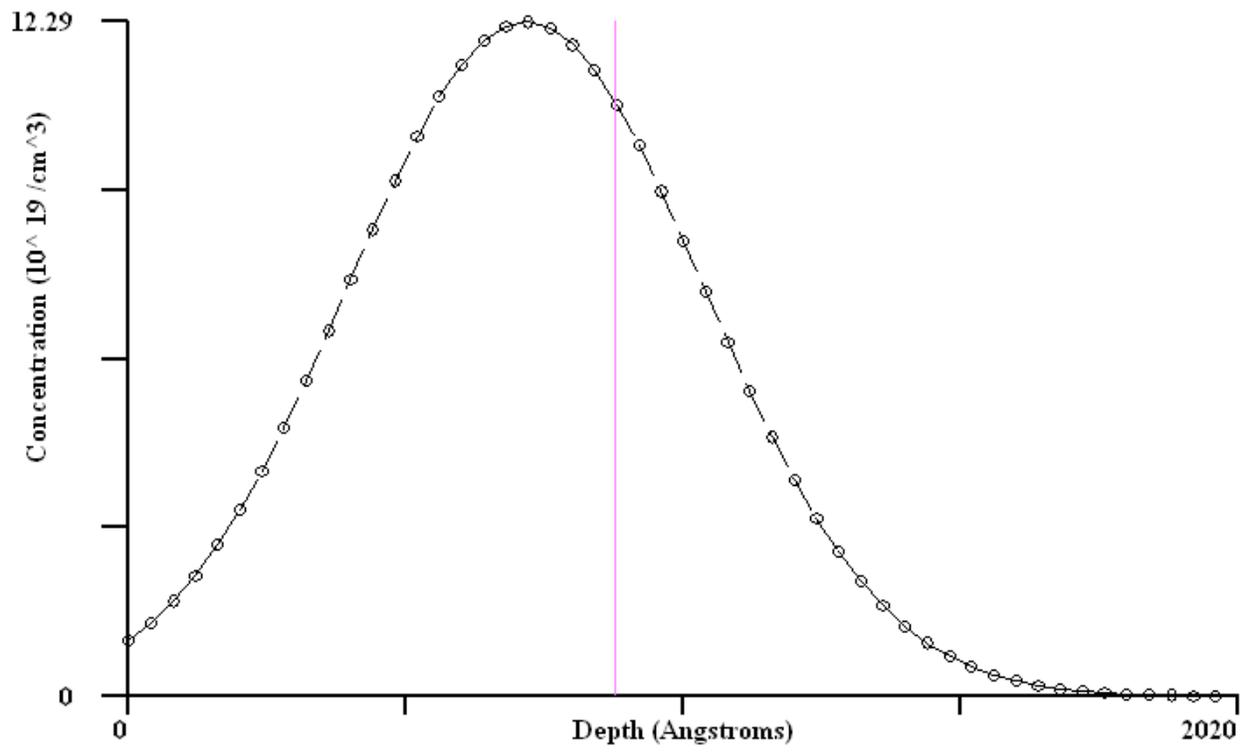


Figure 7a

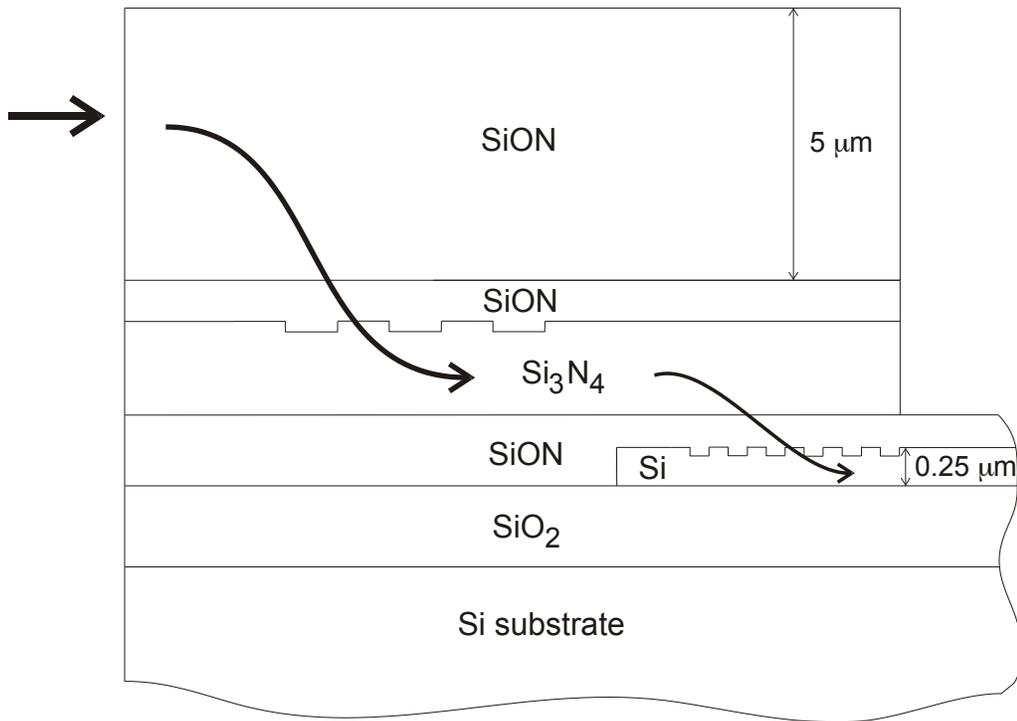


Figure 7b

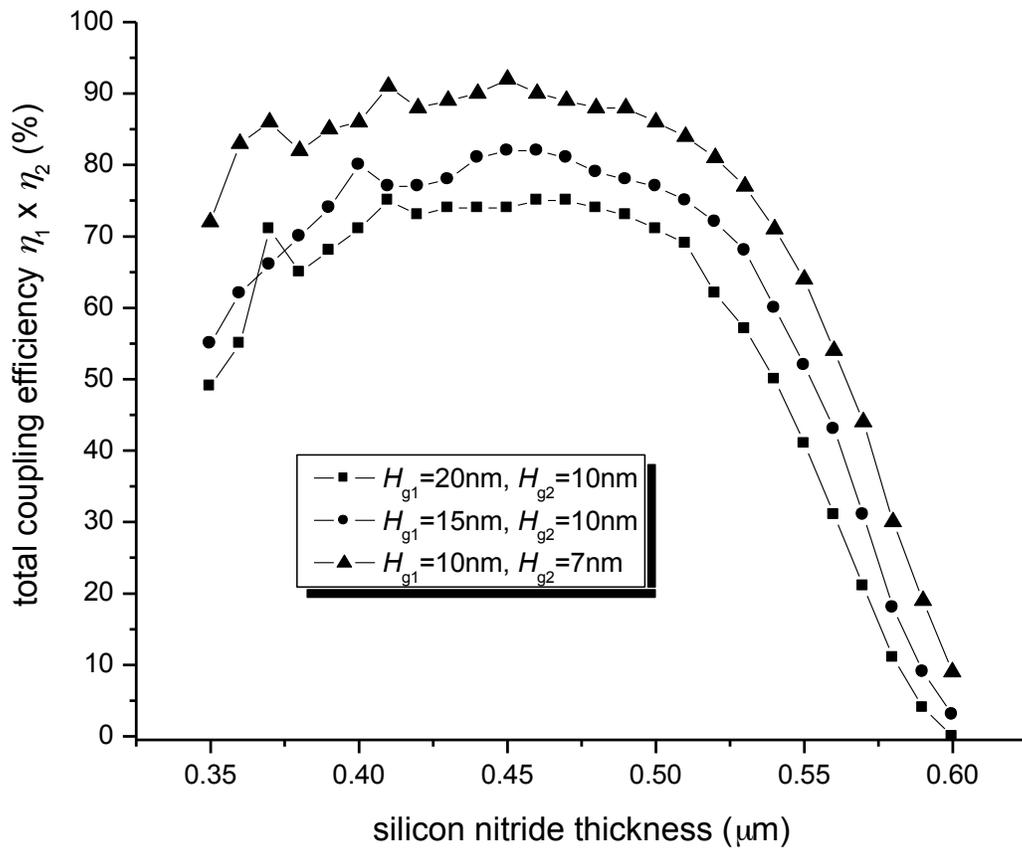


Figure 8a

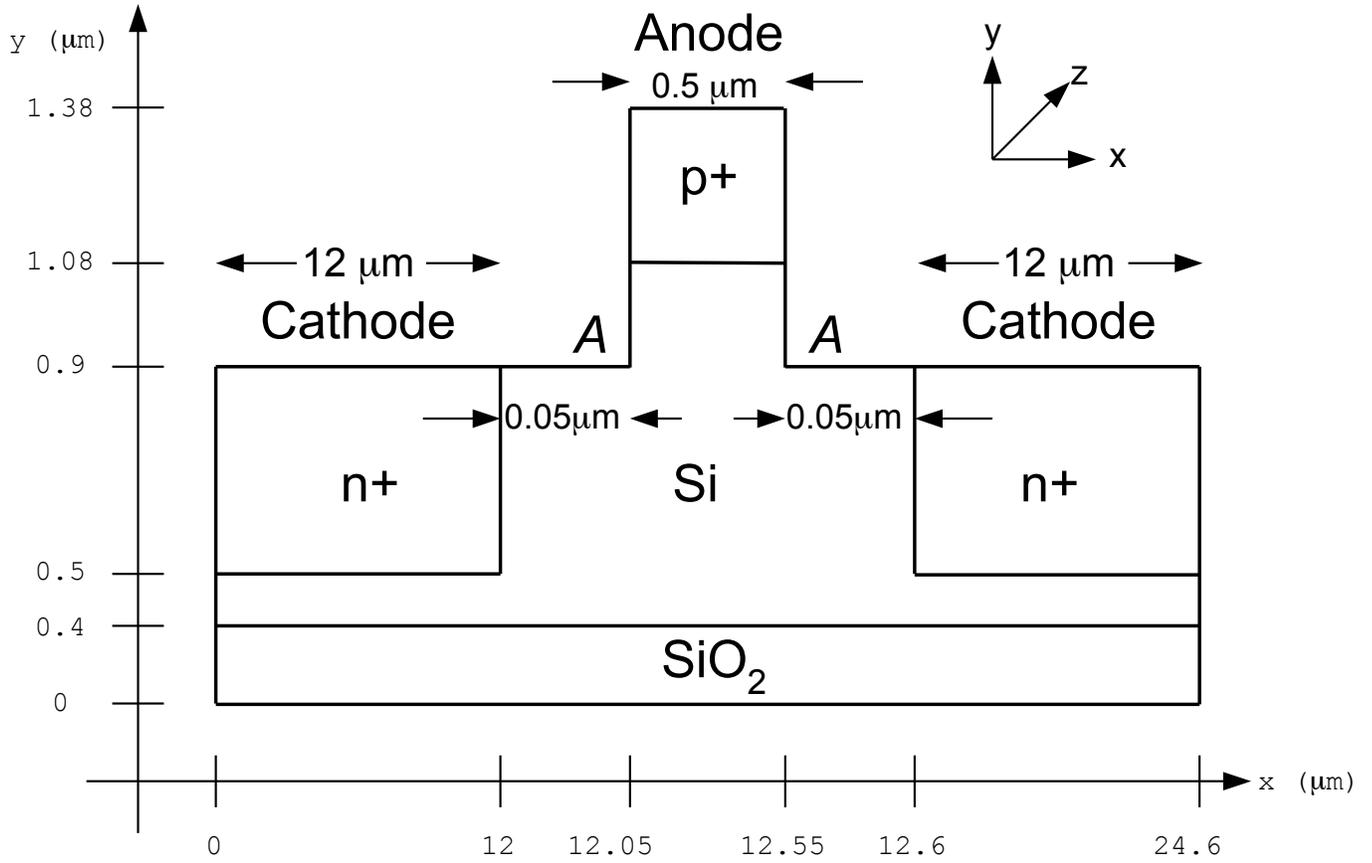


Figure 8b

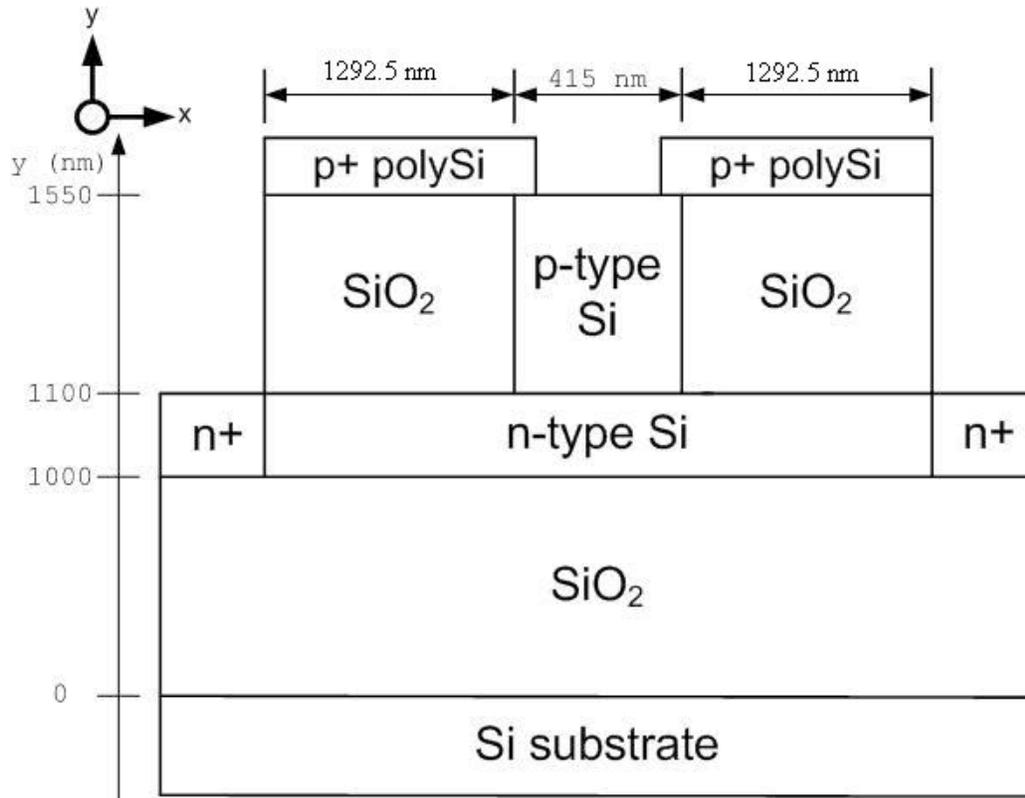


Table 1

<b>Bend Radius</b>	<b>Total circumference</b>	<b>FSR</b>	<b>Q</b>
400 $\mu\text{m}$	3513 $\mu\text{m}$	0.193nm	90,000
300 $\mu\text{m}$	2505 $\mu\text{m}$	0.251nm	170,000
200 $\mu\text{m}$	1877 $\mu\text{m}$	0.333nm	30,000
50 $\mu\text{m}$	934 $\mu\text{m}$	0.691nm	30,000
25 $\mu\text{m}$	777 $\mu\text{m}$	0.815nm	19,200

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**Graham T Reed** joined the University of Surrey in 1989 with the aim of establishing a research activity in guided wave optoelectronics, and now leads an internationally recognised group. He is responsible for initiating a new research field in the UK on Silicon Integrated Optical Circuits, and his group has produced a series of leading technical advances in the field worldwide, notably in optical modulators, grating couplers, and optical sensing applications. A testament to the originality and potential of the silicon work is that Bookham Technology plc adopted it as their core business in the early 1990s. Professor Reed's work is built upon collaborative arrangements with both industry and academia alike. His work has been associated with companies and universities in the UK, the USA, France, Italy, Germany, Japan and Singapore. Professor Reed has contributed to more than 150 publications in the field of guided wave optoelectronics, co-authored the first text book on silicon photonics, and contributed to several patents. He is a Fellow of the IEE.

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**Ching Eng Png** was born in Singapore in 1974. He received the B.Eng. in Electronic and Electrical Engineering (First Class Honours) and the Ph.D. degrees from the University of Surrey, U.K., in 1999 and 2004 respectively. His Ph.D. work was supervised by Prof. G. T. Reed and focused on silicon photonics. From November 1999 to September 2000, he was with Agilent Technologies, Singapore, as a Gigabit Optical Transceiver Engineer. He is currently with the Institute of High Performance Computing, A\*Star, Singapore, working on bio-photonics and silicon photonics. Dr. Png was awarded the Royal Academy of Engineering Prize and the Institution of Electrical Engineers (IEE) Hudswell International Research Scholarship for his work on silicon optical modulators.

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